## **TDA3190**

### COMPLETE TV SOUND CHANNEL

The TDA3190 is a monolithic integrated circuit in a 16-lead dual in-line plastic package. It performs all the functions needed for the TV sound channel :

SGS-THOMSON MICROELECTRONICS

- IF LIMITER AMPLIFIER
- ACTIVE LOW-PASS FILTER
- FM DETECTOR
- DC VOLUME CONTROL
- AF PREAMPLIFIER
- AF OUTPUT STAGE

#### DESCRIPTION

The TDA3190 can give an output power of 4.2 W (d = 10 %) into a 16  $\Omega$  load at Vs = 24 V, or 1.5 W (d = 10 %) into an 8  $\Omega$  load at Vs = 12 V. This performance, together with the FM-IF section characteristics of high sensitivity, high AM rejection and low distortion, enables the device to be used in almost every type of television receivers.

The device has no irradiation problems, hence no external screening is needed.

The TDA3190 is a pin to pin replacement of TDA1190Z.



#### CONNECTION DIAGRAM



#### **BLOCK DIAGRAM**



#### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
Vs	Supply Voltage (pin 10)	28	V
Vi	Input Signal Voltage (pin 1)	1	V
I <sub>o</sub>	Output Peak Current (non-repetitive)	2	A
I <sub>o</sub>	Output Peak Current (repetitive)	1.5	А
Ptot	Power Dissipation : at T <sub>pins</sub> = 90 °C at T <sub>amb</sub> = 70 °C (free air)	4.3 1	W W
T <sub>stg</sub> , T <sub>j</sub>	Storage and Junction Temperature	- 40 to 150	°C

#### THERMAL DATA

	R <sub>th j</sub> pins	Thermal Resistance Junction-pins Thermal Resistance Junction-ambient	Max Max	14 80*	°C/W °C/W
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\* Obtained with the GND pins soldered to printed circuit with minimized copper area.



#### **TEST CIRCUIT**



# **ELECTRICAL CHARACTERISTICS** (refer to the test circuit, V<sub>s</sub> = 24 V, T<sub>amb</sub> = 25 $^{\circ}\text{C}$ unless otherwise specified)

Symbol	Parameter	Test C	onditions	Min.	Typ.	Max.	Unit
Vs	Supply Voltage (pin 14)			9		28	V
Vo	Quiescent Output Voltage (pin 11)	$V_{s} = 24 V$ $V_{s} = 12 V$		11 5.1	12 6	13 6.9	V V
ld	Quiescent Drain Current	$P_1 = 22 \text{ K}\Omega$ $V_s = 24 \text{ V}$ $V_s = 12 \text{ V}$		11	22 19	45 40	mA mA
Po	Output Power	d = 10 % f <sub>o</sub> = 4.5 MHz V <sub>s</sub> = 24 V V <sub>s</sub> = 12 V	$f_{m} = 400 \text{ Hz}$ $\Delta f = \pm 25 \text{ KHz}$ $R_{L} = 16 \Omega$ $R_{L} = 8 \Omega$		4.2 1.5		W
		d = 2 % $f_0 = 4.5 \text{ MHz}$ $V_s = 24 \text{ V}$ $V_s = 12 \text{ V}$	$f_{m} = 400 \text{ Hz}$ $\Delta f = \pm 25 \text{ KHz}$ $R_{L} = 16 \Omega$ $R_{L} = 8 \Omega$		3.5 1.4		w w

Symbol	Parameter	Test Co	onditions	Min.	Тур.	Max.	Unit
V	Input Limiting Voltage (- 3 dB) at Pin 1	$f_0 = 4.5 \text{ MHz}$ $f_m = 400 \text{ Hz}$ $P_1 = 0$	$\Delta f = \pm 7.5 \text{ KHz}$		40	100	μV
d	Distortion	$P_{o} = 50 \text{ mW}$ $f_{o} = 4.5 \text{ MHz}$ $V_{s} = 24 \text{ V}$ $V_{s} = 12 \text{ V}$	$f_{m} = 400 \text{ Hz}$ $\Delta f = \pm 7.5 \text{ KHz}$ $R_{L} = 16 \Omega$ $R_{L} = 8 \Omega$		0.75 1		%
В	Frequency Response of audio- amplifier (- 3 dB)	$R_{L} = 16 \Omega$ $C_{7} = 470 \text{ pF}$ $R_{1} = 82 \Omega$ $R_{1} = 47 \Omega$	C <sub>8</sub> = 120 pF P <sub>1</sub> = 22 KΩ		70 to 1200 70 to 7000		Hz Hz
Vo	Recovered Audio Voltage (pin 16)	$V_i \ge 1 \text{ mV}$ $f_m = 400 \text{ Hz}$ $P_1 = 0$	$f_o = 4.5 \text{ MHz}$ $\Delta f = \pm 7.5 \text{ KHz}$		120		mV
AMR	Ampliture Modulation Rejection	$V_i \ge 1 \text{ mV}$ $f_m = 400 \text{ Hz}$ $m = 0.3$	$f_o = 4.5 \text{ MHz}$ $\Delta_f = \pm 25 \text{ KHz}$		55		dB
S + N N	Signal to Noise Ratio	$V_{I} \ge 1 \text{ mV}$ $f_{o} = 4.5 \text{ MHz}$ $\Delta f = \pm 25 \text{ KHz}$	V <sub>o</sub> = 4 V f <sub>m</sub> = 400 Hz	50	65	-	dB
R <sub>3</sub>	External Feedback Resistance (between pins 9 and 11)					25	KΩ
Ri	Input Resistance (pin 1)	$V_i = 1 \text{ mV}$			30		KΩ
Ci	Input Capacitance (pin 1)	fo = 4.5 MHz			5		pF
SVR	Supply Voltage Rejection	$ \begin{array}{l} R_{L} = 16 \ \Omega \\ f_{ripple} = 120 \ Hz \\ P_1 = 22 \ K\Omega \end{array} $			46		dB
Av	DC Volume Control Attenuation	P <sub>1</sub> = 12 KΩ			90		dB

#### **ELECTRICAL CHARACTERISTICS** (continued)











Figure 3 : Amplitude Modulation Rejection vs. Input Signal.



Figure 5 : Recovered Audio Voltage vs. Unloaded Q Factor of the Detector Coil.



Figure 7 : Distortion vs. Frequency Deviation.



Figure 4 :  $\triangle$  AMR vs. Tuning Frequency Change.



Figure 6 : Distortion vs. Output Power.



Figure 8 : Distortion vs. Tunning Frequency Change.





Figure 9 : Audio Amplifier Frequency Response.



Figure 11 : Supply Voltage Ripple Rejection vs; Volume Control Attenuation.



Figure 13 : Maximum Power Dissipation vs. Supply Voltage (sine wave operation).



Figure 10 : Supply Voltage Ripple Rejection vs.



Figure 12 : Output Power vs. Supply Voltage.



Figure 14 : Power Dissipation and Efficiency vs. Output Power.





0 5 10 15 20 25 30 V<sub>5</sub>(V)

Figure 15 : Quiescent Output Voltage (pin 11) vs. Supply Voltage.

#### **APPLICATION INFORMATION**

The electrical characteristics of the TDA3190 remain almost constant over the frequency range 4.5 to 6 MHz, therefore it can be used in all television standards (FM mod.). The TDA3190 has a high input impedance, so it can work with a ceramic filter or with a tuned circuit that provide the necessary input selectivity.

The value of the resistors connected to pin 9, determine the AC gain of the audio frequency amplifier. This enables the desired gain to be selected in relation to the frequency deviation at which the output stage of the AF amplifier, must enter into clipping. Capacitor C8, connected between pins 10 and 11, determines the upper cutoff frequency of the audio bandwidth. To increase the bandwidth the values of C8 and C7 must be reduced, keeping the ratio C7/C8 as shown in the table of fig. 16.

The capacitor connected between pin 16 and ground, together with the internal resistor of 10 K $\Omega$  forms the de-emphasis network. The Boucherot cell eliminates the high frequency oscillations caused by the inductive load and the wires connecting the loud-speaker.



Figure 16 : Typical Application Circuit.

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Figure 17 : P.C	. Board and Comp	onent Layout of the	Circuit Shown in Fig	. 16 (1 : 1 scale).
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#### MOUNTING INSTRUCTION

The Rth j-amb of the TDA3190 can be reduced by soldering the GND pins to a suitable copper area of the printed circuit board (fig. 18) or to an external heatsink (fig. 19).

The diagram of figure 20 shows the maximum dissipable power Ptot and the Rth j-amb as a function of the side "I" of two equal square copper areas having a thickness of  $35 \,\mu$  (1.4 mils).

During soldering the pins temperature must not exceed 260 °C and the soldering time must not be longer than 12 seconds.

The external heatsink or printed circuit copper area must be connected to electrical ground.





Figure 18 : Example of P.C. Board Copper Area which is used as Heatsink.

Figure 19 : External Heatsink Mounting Example.



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Figure 20 : Maximum Dissipable Power and Junction to Ambient Thermal Resistance vs. Side "I"



Figure 21 : Maximum Allowable Power Dissipation vs. Ambient Temperature.



