

## TDA1023 Time-Proportional Triac Trigger

### Product Specification

#### Linear Products

#### DESCRIPTION

The TDA1023 is a bipolar integrated circuit for controlling triacs in the time-proportional or burst firing mode. It permits very precise temperature control of heating equipment and is especially suited to the control of panel heaters. The circuit generates positive-going trigger pulses and complies with the regulations on radio interference and mains distortion.

#### FEATURES

- Adjustable proportional range width
- Adjustable hysteresis
- Adjustable trigger pulse width
- Adjustable firing burst repetition time
- Control range translation facility
- Failsafe operation
- Supplied from the mains
- Provides supply for external temperature bridge

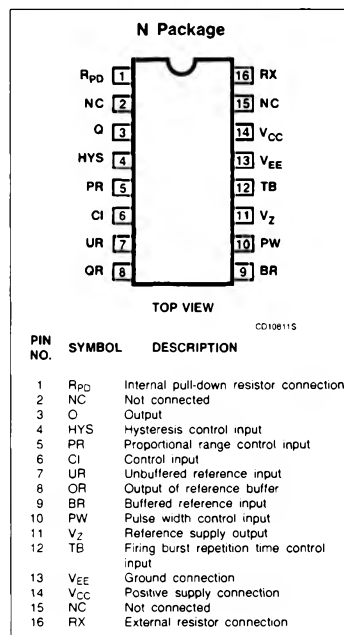
#### APPLICATIONS

- Triac control
- Temperature control
- Panel heater control

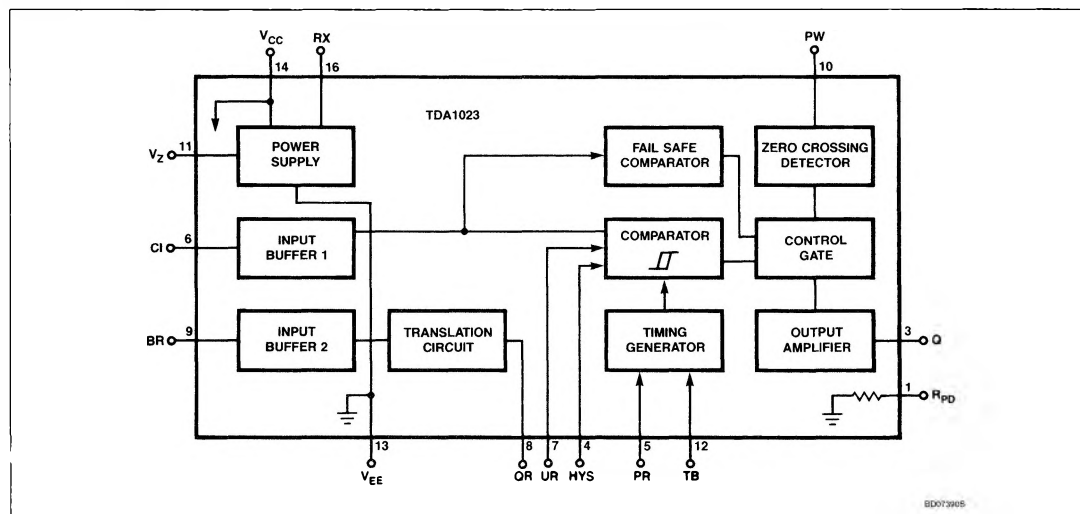
#### ORDERING INFORMATION

DESCRIPTION	TEMPERATURE RANGE	ORDER CODE
16-Pin Plastic DIP (SOT-38)	-20°C to +75°C	TDA1023N

#### PIN CONFIGURATION



#### BLOCK DIAGRAM



## Time-Proportional Triac Trigger

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## ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	RATING	UNIT
$V_{CC}$	Supply voltage, DC	16	V
$I_{16(AV)}$ $I_{16(RM)}$ $I_{16(SM)}$	Supply current average repetitive peak non-repetitive peak	30 100 2	mA mA A
$V_I$	Input voltage, all inputs	16	V
$I_{6, 7, 9, 10}$	Input current, CI, UR, BR, PW input	10	mA
V	Voltage on $R_{PD}$ connection (Pin 1)	16	V
$V_{3, 8, 11}$	Output voltage, Q, QR, $V_Z$ output	16	V
$-I_{OH(AV)}$ $-I_{OH(M)}$	Output current average peak, max. 300 $\mu$ s	30 700	mA mA
$P_{TOT}$	Total power dissipation	500	mW
$T_{STG}$	Storage temperature range	-65 to +150	$^{\circ}$ C
$T_A$	Operating ambient temperature range	-20 to +75	$^{\circ}$ C

DC ELECTRICAL CHARACTERISTICS  $V_{CC} = 11$  to 16V;  $T_A = -20^{\circ}$ C to  $+75^{\circ}$ C, unless otherwise specified.

SYMBOL	PARAMETER	LIMITS			UNIT
		Min	Typ	Max	
Supply: $V_{CC}$ and RX (Pins 14 and 16)					
$V_{CC}$ $\Delta V_{CC}/\Delta I_{16}$	Internally stabilized supply voltage at $I_{16} = 10\text{mA}$ Variation with $I_{16}$	12	13.7 30	15	V mV/mA
$I_{16}$ $I_{16}$	Supply current at $V_{16-13} = 11$ to 16V; $I_{10} = 1\text{mA}$ ; $f = 50\text{Hz}$ ; Pin 11 open; $V_{6-13} > V_{7-13}$ ; Pins 4 and 5 open Pins 4 and 5 grounded			6 7.1	mA mA
Reference supply output $V_Z$ for external temperature bridge 0(Pin 11)					
$V_{11-13}$	Output voltage		8		V
$-I_{11}$	Output current			1	mA
Control and reference inputs CI, BR, and UR (Pins 6, 9, and 7)					
$V_{6-13}$	Input voltage to inhibit the output		7.6		V
$I_6, 7, 9$	Input current at $V_I = 4\text{V}$			2	$\mu\text{A}$
Hysteresis control input HYS (Pin 4)					
$\Delta V_6$ $\Delta V_6$	Hysteresis, Pin 4 open Pin 4 grounded	9	20 320	40	mV mV
Proportional range control input PR (Pin 5)					
$\Delta V_6$ $\Delta V_6$	Proportional range, Pin 5 open Pin 5 grounded	50	80 400	130	mV mV
Pulse width control input PW (Pin 10)					
$t_W$	Pulse width at $I_{10(\text{RMS})} = 1\text{mA}$ ; $f = 50\text{Hz}$	100	200	300	$\mu\text{s}$
Firing burst repetition time control input $t_B$ (Pin 12)					
$t_B/C_T$	Firing burst repetition time, ratio to capacitor $C_T$	320	600	960	ms/ $\mu\text{F}$

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SYMBOL	PARAMETER	LIMITS			UNIT
		Min	Typ	Max	
Output of reference buffer QR (Pin 8)					
$V_{8-13}$	Output voltage				
$V_{8-13}$	at input voltage $V_{9-13} = 1.6V$		3.2		V
$V_{8-13}$	$V_{9-13} = 4.8V$		4.8		V
$V_{8-13}$	$V_{9-13} = 8V$		6.4		V
Output Q (Pin 3)					
$V_{OH}$	Output voltage HIGH at $-I_{OH} = 150mA$	10			V
$-I_{OH}$	Output current HIGH			150	mA
Internal pull-down resistor $R_{PD}$ (Pin 1)					
$R_{PD}$	Resistance to $V_{EE}$	1	1.5	3	k $\Omega$

**FUNCTIONAL DESCRIPTION**

The TDA1023 generates pulses to trigger a triac. These trigger pulses coincide with the zero crossings of the mains voltage. This minimizes RF interference and transients on the mains supply. The trigger pulses come in bursts, with the net effect that the load is periodically switched on and off. This further minimizes mains pollution. The average power in the load is varied by varying the duration of the trigger pulse burst in accordance with the voltage difference between the control input CI and the reference input, either UR or BR.

**Power Supply:  $V_{CC}$ , RX and  $V_Z$  (Pins 14, 16, and 11)**

The TDA1023 is supplied from the AC mains via resistor  $R_D$  to the RX connection (Pin 16), while the  $V_{EE}$  connection (Pin 13) is connected to the neutral line (see Figure 5). A smoothing capacitor  $C_S$  has to be connected between the  $V_{CC}$  and  $V_{EE}$  connections.

The circuit contains a string of stabilizer diodes between the RX and  $V_{EE}$  connections that limit the DC supply voltage and a rectifier diode between the RX and  $V_{CC}$  connections (see Figure 1).

At Pin 11 the device provides a stabilized reference voltage  $V_Z$  for an external temperature sensing bridge.

The operation of the supply arrangement is as follows. During the positive half of the mains cycles the current through external voltage dropping resistor  $R_D$  charges the external smoothing capacitor  $C_S$  until RX reaches the stabilizing voltage of the internal stabilizer diodes.  $R_D$  should be chosen such that it can supply the current  $I_{CC}$  for the TDA1023 itself plus the average output current  $I_{3(AV)}$  plus the current required from the  $V_Z$  connection for an external temperature bridge, and recharge the smoothing capacitor  $C_S$  (see Figures 7 to 10). Any excess current

is bypassed by the internal stabilizer diodes. Note that the maximum rated supply current must not be exceeded.

During the negative half of the mains cycles, external smoothing capacitor  $C_S$  has to supply the sum of the currents mentioned above. Its capacitance must be high enough to maintain the supply voltage above the minimum specified limit.

Dissipation in resistor  $R_D$  is halved by connecting a diode in series (see Figures 2 and 7 to 10).

A further reduction of dissipation is possible by using a high-quality voltage dropping capacitor  $C_D$  in series with a resistor  $R_{SD}$  (see Figures 2 and 12). A suitable VDR connected across the mains provides protection of the TDA1023 and of the triac against mains-borne transients.

**Control and Reference Inputs CI, BR and UR (Pins 6, 9, and 7)**

For room temperature control ( $5^\circ C$  to  $30^\circ C$ ) the best performance is obtained by using the translation circuit. The buffered reference input BR (Pin 9) is used as a reference input, and the output of the reference buffer QR (Pin 8) is connected to the unbuffered reference input UR (Pin 7). In this arrangement, the translation circuit ensures that most of the potentiometer rotation can be used to cover the room temperature range. This provides an accurate temperature setting and a linear temperature scale.

If the translation circuit is not required, the unbuffered reference input UR (Pin 7) is used as a reference input. The buffered reference input BR (Pin 9) must be connected to the reference supply output  $V_Z$  (Pin 11).

For proportional power control the unbuffered reference input UR (Pin 7) must be connected to the firing burst repetition time control input

TB (Pin 12), and the buffered reference input BR (Pin 9), which is inactive now, must be connected to the reference supply output  $V_Z$  (Pin 11).

In all arrangements, the train of output pulses becomes longer when the voltage at the control input CI (Pin 6) becomes lower.

**Proportional Range Control Input PR (Pin 5)**

With the proportional range control input PR open, the output duty factor changes from 0% to 100% by a variation of 80mV at the control input CI (Pin 6). For temperature control, this corresponds with a temperature difference of only 1k.

This range may be increased to 400mV, i.e., 5k, by connecting the proportional range control input PR (Pin 5) to ground. Intermediate values are obtained by connecting the PR input to ground via a resistor  $R_5$  (see Table 1).

**Hysteresis Control Input HYS (Pin 4)**

With the hysteresis control input HYS (Pin 4) open, the device has a built-in hysteresis of 20mV. For temperature control this corresponds with 0.25k.

Hysteresis is increased to 320mV, corresponding with 4k, by grounding HYS (Pin 4). Intermediate values are obtained by connecting Pin 4 to ground via a resistor  $R_4$ . See Table 1 for a set of values for  $R_4$  and  $R_5$  giving a fixed ratio between hysteresis and proportional range.

**Trigger Pulse Width Control Input PW (Pin 10)**

The trigger pulse width may be adjusted to the value required for the triac by choosing the value of the external synchronization resistor  $R_S$  between the trigger pulse width control input PW (Pin 10) and the AC mains.

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The pulse width is inversely proportional to the input current (see Figure 11).

**Output Q (Pin 3)**

Since the circuit has an open-emitter output, it is capable of sourcing current; i.e., supplying a current out of the output. Therefore, it is especially suited for generating positive-going trigger pulses. The output is current-limited

and protected against short-circuits. The maximum output current is 150mA and the output pulses are stabilized at 10V for output currents up to that value.

A gate resistor  $R_G$  must be connected between the output Q and the triac gate to limit the output current to the minimum required by the triac (see Figures 3 to 6). This minimizes

the total supply current and the power dissipation.

**Pull-Down Resistor  $R_{PD}$  (Pin 1)**

The TDA1023 includes a  $1.5k\Omega$  pull-down resistor  $R_{PD}$  between Pins 1 and 13 ( $V_{EE}$ , ground connection), intended for use with sensitive triacs.

**Table 1. Adjustment of Proportional Range and Hysteresis.  
Combinations of Resistor Values Giving Hysteresis  
>  $\frac{1}{4}$  Proportional Range**

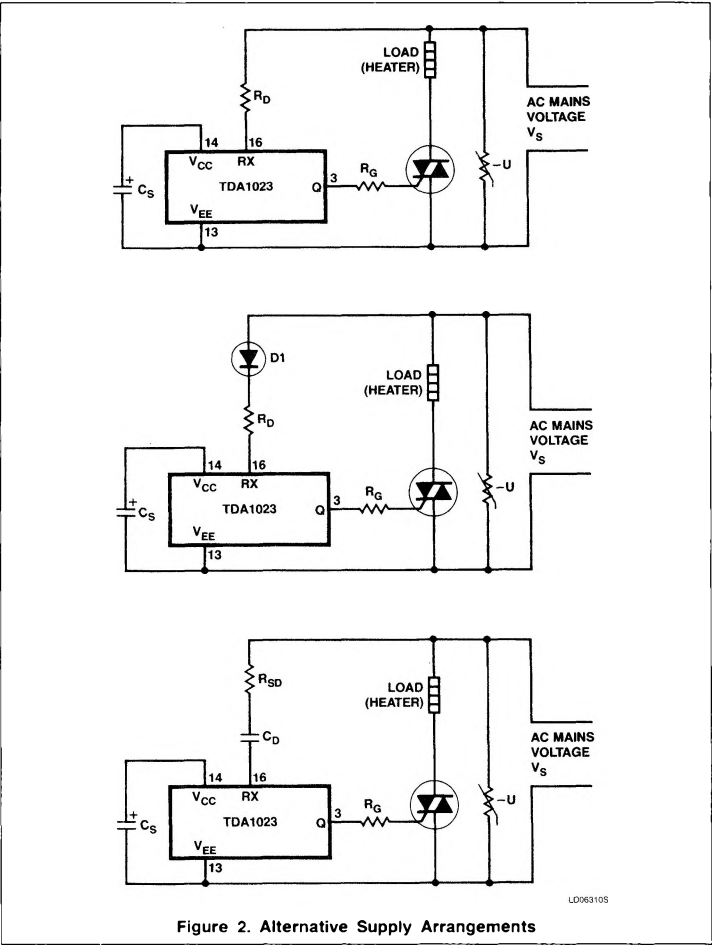
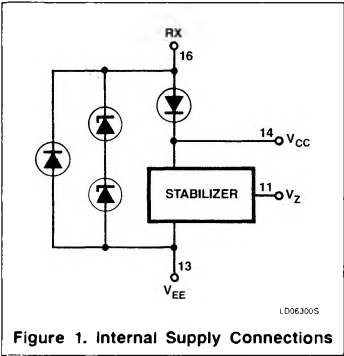
PROPORTIONAL RANGE (mV)	PROPORTIONAL RANGE RESISTOR $R_5$ ( $k\Omega$ )	MINIMUM HYSTERESIS (mV)	MAXIMUM HYSTERESIS RESISTOR $R_4$ ( $k\Omega$ )
80	Open	20	Open
160	3.3	40	9.1
240	1.1	60	4.3
320	0.43	80	2.7
400	0	100	1.8

**Table 2. Timing Capacitor  $C_T$  Values (Electrolytic Capacitors)**

EFFECTIVE DC VALUE ( $\mu F$ )	MARKED AC SPECIFICATION	
	$\mu F$	V
68	47	25
47	33	40
33	22	25
22	15	40
15	10	25
10	6.8	40

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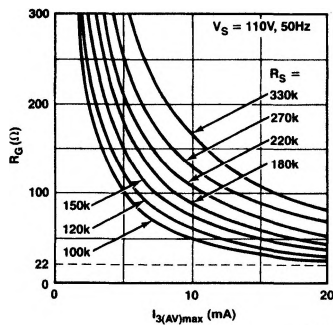


Figure 3

CP088505

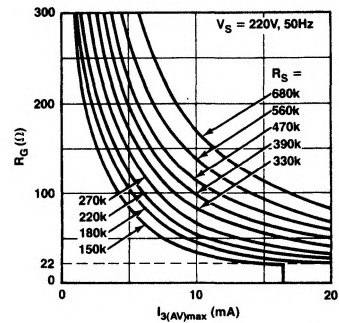


Figure 4

CP088605

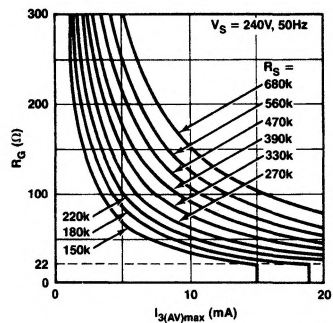


Figure 5

CP088705

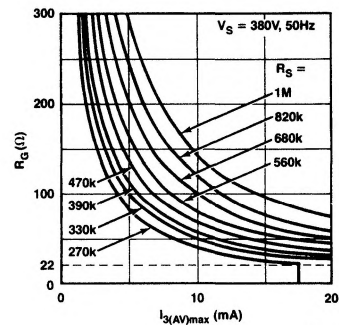


Figure 6

CP088805

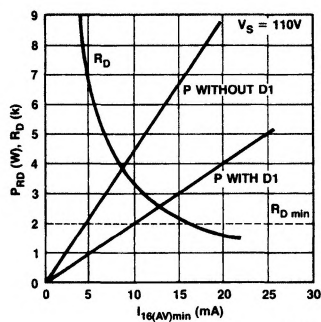


Figure 7

CP088905

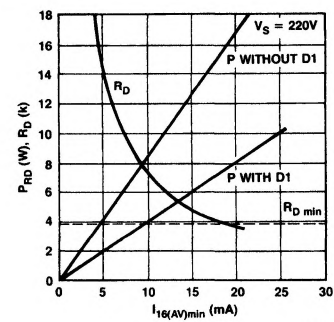


Figure 8

CP089005

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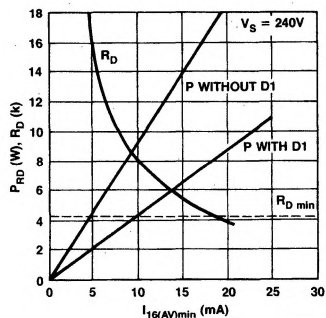


Figure 9

OP089105

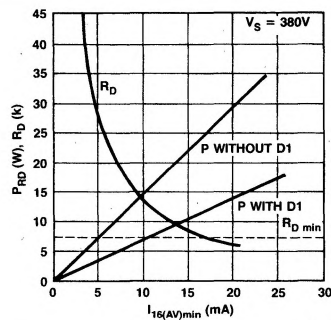


Figure 10

OP089205

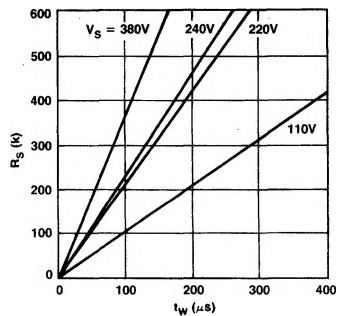


Figure 11

OP089305

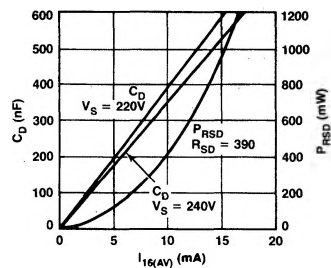


Figure 12. Nominal Value of Voltage Dropping Capacitor  $C_D$  and Power  $P_{RSD}$  Dissipated in Voltage Dropping Resistor  $R_{SD}$  as a Function of the Average Supply Current  $I_{16(AV)}$  with the Mains Supply Voltage  $V_S$  as a Parameter

OP089405

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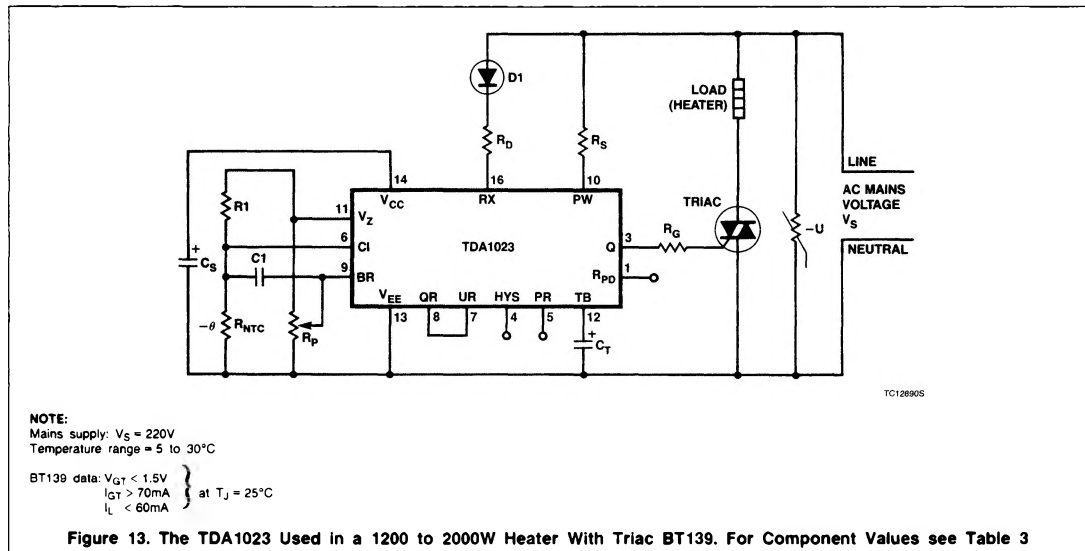


Table 3. Temperature Controller Component Values (see Figure 13)

SYMBOL	PARAMETER	VALUE	REMARKS
$t_W$	Trigger pulse width	75 $\mu s$	See BT139 data sheet
$R_S$	Synchronization resistor	180k $\Omega$	See Figure 11
$R_G$	Gate resistor	110 $\Omega$	See Figure 4
$I_{3(AV)}$	Maximum average gate current	4.1mA	See Figure 6
$R_4$	Hysteresis resistor	NC	See Table 1
$R_5$	Proportional band resistor	NC	See Table 1
$I_{16(AV)}$	Minimum required supply current	11.1mA	
$R_D$	Mains dropping resistor	6.2k $\Omega$	See Figure 8
$P_{RD}$	Power dissipated in $R_D$	4.6W	See Figure 8
$C_T$	Timing capacitor (eff. value)	68 $\mu F$	See Table 2
VDR	Voltage-dependent resistor	250V <sub>AC</sub>	Cat. no. 2322 593 62512
D1	Rectifier diode	BYW56	
$R_1$	Resistor to Pin 11	18.7k $\Omega$	1% tolerance
$R_{NTC}$	NTC thermistor (at 25°C)	22k $\Omega$	B = 4200k Cat. no. 2322 642 12223
$R_P$	Potentiometer	22k $\Omega$	
$C_1$	Capacitor between Pins 6 and 9	47nF	
$C_S$	Smoothing capacitor	220 $\mu F$ ; 16V	
<b>If <math>R_D</math> and D1 are replaced by <math>C_D</math> and <math>R_{SD}</math></b>			
$C_D$	Mains dropping capacitor	470nF	}
$R_{SD}$	Series dropping resistor	390 $\Omega$	
$P_{RSD}$	Power dissipated in $R_{SD}$	0.6W	
VDR	Voltage-dependent resistor	250V <sub>AC</sub>	Cat. no. 2322 594 62512