

LM4671 Boomer® Audio Power Amplifier Series Filterless High Efficiency 2.5W Switching Audio Amplifier

Check for Samples: [LM4671](#)

FEATURES

- No Output Filter Required for Inductive Loads
- Externally Configurable Gain
- Very Fast Turn On Time: 17µs (typ)
- Minimum External Components
- "Click and pop" Suppression Circuitry
- Micro-Power Shutdown Mode
- Available in Space-Saving DSBGA Package

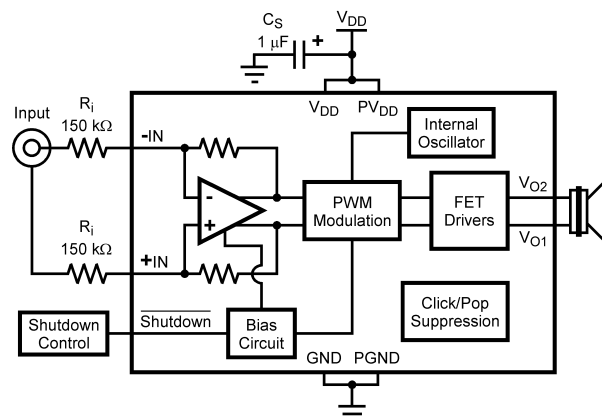
APPLICATIONS

- Mobile Phones
- PDAs
- Portable Electronic Devices

KEY SPECIFICATIONS

- Efficiency @ 3.6V, 100mW into 8Ω speaker: 80 % (typ)
- Efficiency @ 3.6V, 400mW into 8Ω speaker: 88 % (typ)
- Efficiency @ 5V, 1W into 8Ω speaker: 86 % (typ)
- Quiescent current, 3.6V supply: 2.8 mA (typ)
- Total shutdown power supply current: 0.01 µA (typ)
- Single supply range: 2.4 to 5.5 V

Typical Application


Figure 1. Typical Audio Amplifier Application Circuit


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Connection Diagram

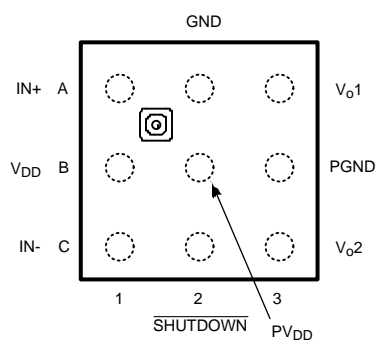


Figure 2. 9 Bump DSBGA Package Top View
See Package Number YZR0009AAA



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings⁽¹⁾⁽²⁾⁽³⁾

Supply Voltage ⁽¹⁾		6.0V
Storage Temperature		-65°C to +150°C
Voltage at Any Input Pin		$V_{DD} + 0.3V \geq V \geq GND - 0.3V$
Power Dissipation ⁽⁴⁾		Internally Limited
ESD Susceptibility, all other pins ⁽⁵⁾		2.0kV
ESD Susceptibility ⁽⁶⁾		200V
Junction Temperature (T_{JMAX})		150°C
Thermal Resistance	θ_{JA} (DSBGA)	220°C/W
Soldering Information		
See AN-1112 "microSMD Wafers Level Chip Scale Package."		

- (1) All voltages are measured with respect to the ground pin, unless otherwise specified.
- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not specified for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (4) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4671, $T_{JMAX} = 150^\circ\text{C}$. The typical θ_{JA} is 220°C/W for the DSBGA package.
- (5) Human body model, 100pF discharged through a 1.5kΩ resistor.
- (6) Machine Model, 220pF–240pF discharged through all pins.

Operating Ratings⁽¹⁾⁽²⁾

Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$	$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$
Supply Voltage		$2.4V \leq V_{DD} \leq 5.5V$

- (1) All voltages are measured with respect to the ground pin, unless otherwise specified.
- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not specified for parameters where no limit is given, however, the typical value is a good indication of device performance.

Electrical Characteristics⁽¹⁾⁽²⁾

The following specifications apply for $A_V = 2V/V$ ($R_I = 150\text{k}\Omega$), $R_L = 15\mu\text{H} + 8\Omega + 15\mu\text{H}$ unless otherwise specified. Limits apply for $T_A = 25^\circ\text{C}$.

Symbol	Parameter	Conditions	LM4671		Units (Limits)
			Typical ⁽³⁾	Limit ^{(4) (5)}	
$ V_{OS} $	Differential Output Offset Voltage	$V_I = 0V$, $A_V = 2V/V$, $V_{DD} = 2.4V$ to $5.0V$	5		mV (max)
$PSRR_{GSM}$	GSM Power Supply Rejection Ratio	$V_{DD} = 2.4V$ to $5.0V$	61		dB (min)
$CMRR_{GSM}$	GSM Common Mode Rejection Ratio	$V_{DD} = 2.4V$ to $5.0V$ $V_{IC} = V_{DD}/2$ to $0.5V$, $V_{IC} = V_{DD}/2$ to $V_{DD} - 0.8V$	68		dB (min)
$ I_{IH} $	Logic High Input Current	$V_{DD} = 5.0V$, $V_I = 5.5V$	17	100	μA (max)
$ I_{IL} $	Logic Low Input Current	$V_{DD} = 5.0V$, $V_I = -0.3V$	0.9	5	μA (max)

- (1) All voltages are measured with respect to the ground pin, unless otherwise specified.
- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not specified for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) Typical specifications are specified at 25°C and represent the parametric norm.
- (4) Tested limits are specified to AOQL (Average Outgoing Quality Level).
- (5) Datasheet min/max specification limits are specified by design, test, or statistical analysis.

Electrical Characteristics⁽¹⁾⁽²⁾ (continued)

The following specifications apply for $A_V = 2V/V$ ($R_I = 150k\Omega$), $R_L = 15\mu H + 8\Omega + 15\mu H$ unless otherwise specified. Limits apply for $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4671		Units (Limits)
			Typical ⁽³⁾	Limit ^{(4) (5)}	
I_{DD}	Quiescent Power Supply Current	$V_{IN} = 0V$, No Load, $V_{DD} = 5.0V$	6.4		mA (max)
		$V_{IN} = 0V$, No Load, $V_{DD} = 3.6V$	3.8	6.2	mA
		$V_{IN} = 0V$, No Load, $V_{DD} = 2.4V$	2.0	3.0	mA (max)
I_{SD}	Shutdown Current	$V_{SHUTDOWN} = 0V$ $V_{DD} = 2.4V$ to $5.0V$	0.01	1	μA (max)
V_{SDIH}	Shutdown voltage input high		1.2	1.4	V (min)
V_{SDIL}	Shutdown voltage input low		1.1	0.4	V (max)
R_{OSD}	Output Impedance	$V_{SHUTDOWN} = 0.4V$	100		k Ω
A_V	Gain		$300k\Omega/R_I$	$270k\Omega/R_I$ $330k\Omega/R_I$	V/V (min) V/V (max)
R_{SD}	Resistance from Shutdown Pin to GND		300		k Ω
P_O	Output Power	$R_L = 15\mu H + 4\Omega + 15\mu H$ THD = 10% (max) $f = 1kHz$, 22kHz BW $V_{DD} = 5V$ $V_{DD} = 3.6V$ $V_{DD} = 2.5V$	2.5 1.3 520		W W mW
		$R_L = 15\mu H + 4\Omega + 15\mu H$ THD = 1% (max) $f = 1kHz$, 22kHz BW $V_{DD} = 5V$ $V_{DD} = 3.6V$ $V_{DD} = 2.5V$	2.21 1.06 420		W W mW
P_O	Output Power	$R_L = 15\mu H + 8\Omega + 15\mu H$ THD = 10% (max) $f = 1kHz$, 22kHz BW			
		$V_{DD} = 5V$	1.7		W
		$V_{DD} = 3.6V$	870		mW
		$V_{DD} = 2.5V$	425		mW
		$R_L = 15\mu H + 8\Omega + 15\mu H$ THD = 1% (max) $f = 1kHz$, 22kHz BW			
		$V_{DD} = 5V$	1.19		W
		$V_{DD} = 3.6V$	700	600	mW
$V_{DD} = 2.5V$	350		mW		
THD+N	Total Harmonic Distortion + Noise	$V_{DD} = 5V$, $P_O = 0.1W_{RMS}$, $f = 1kHz$	0.09		%
		$V_{DD} = 3.6V$, $P_O = 0.1W_{RMS}$, $f = 1kHz$	0.04		%
		$V_{DD} = 3.6V$, $P_O = 0.1W_{RMS}$, $f = 5kHz$	0.12		%
		$V_{DD} = 3.6V$, $P_O = 0.1W_{RMS}$, $f = 10kHz$	0.05		%

Electrical Characteristics⁽¹⁾⁽²⁾ (continued)

The following specifications apply for $A_V = 2V/V$ ($R_I = 150k\Omega$), $R_L = 15\mu H + 8\Omega + 15\mu H$ unless otherwise specified. Limits apply for $T_A = 25^\circ C$.

Symbol	Parameter	Conditions	LM4671		Units (Limits)
			Typical ⁽³⁾	Limit ^{(4) (5)}	
PSRR	Power Supply Rejection Ratio	$V_{DD} = 3.6V, 5V$ $V_{Ripple} = 200mV_{PP}$ Sine, $f_{Ripple} = 217Hz$ Inputs to AC GND, $C_I = 2\mu F$	61.8		dB
		$V_{DD} = 3.6V, 5V$ $V_{Ripple} = 200mV_{PP}$ Sine, $f_{Ripple} = 1kHz$ Inputs to AC GND, $C_I = 2\mu F$	59.8		dB
		$V_{DD} = 3.6V, 5V$ $V_{Ripple} = 200mV_{PP}$ Sine, $f_{Ripple} = 10kHz$ Inputs to AC GND, $C_I = 2\mu F$	48.7		dB
		$V_{DD} = 3.6V, 5V$ $V_{Ripple} = 200mV_{PP}$ Sine, $f_{Ripple} = 217Hz$ $f_{IN} = 1kHz, P_O = 10mW_{RMS}$	65.7		dB
SNR	Signal to Noise Ratio	$V_{DD} = 5V, P_O = 1W_{RMS}$	93		dB
ϵ_{OUT}	Output Noise	$V_{DD} = 3.6V, f = 20Hz - 20kHz$ Inputs to AC GND, $C_I = 2\mu F$ No Weighting	58		μV_{RMS}
		$V_{DD} = 3.6V$, Inputs to AC GND $C_I = 2\mu F$, A Weighted	38		μV_{RMS}
CMRR	Common Mode Rejection Ratio	$V_{DD} = 3.6V, V_{Ripple} = 1V_{PP}$ Sine $f_{Ripple} = 217Hz$	68.3		dB
T_{WU}	Wake-up Time	$V_{DD} = 3.6V$	17	49	μs (max)
T_{SD}	Shutdown Time		140		μs

External Components Description

(Figure 1)

Components		Functional Description
1.	C_S	Supply bypass capacitor which provides power supply filtering. Refer to POWER SUPPLY BYPASSING for information concerning proper placement and selection of the supply bypass capacitor.
2.	C_I	Input AC coupling capacitor which blocks the DC voltage at the amplifier's input terminals.

Typical Performance Characteristics

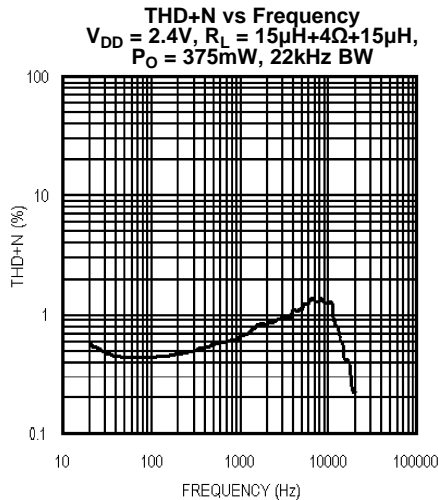


Figure 3.

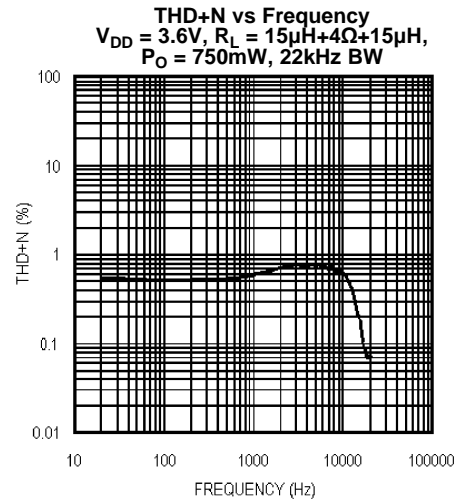


Figure 4.

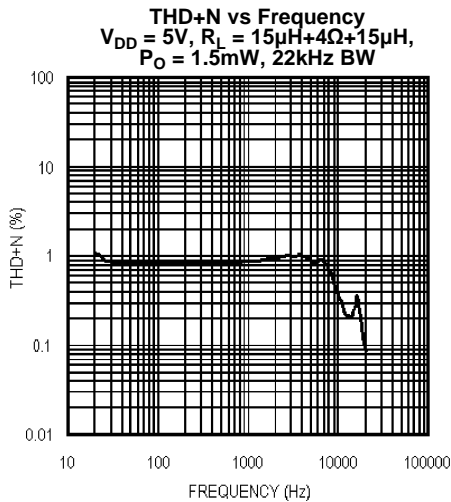


Figure 5.

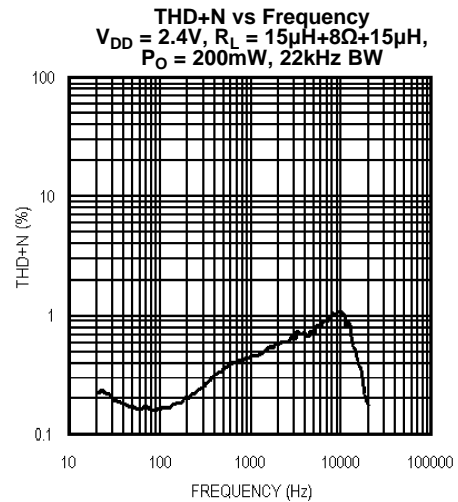


Figure 6.

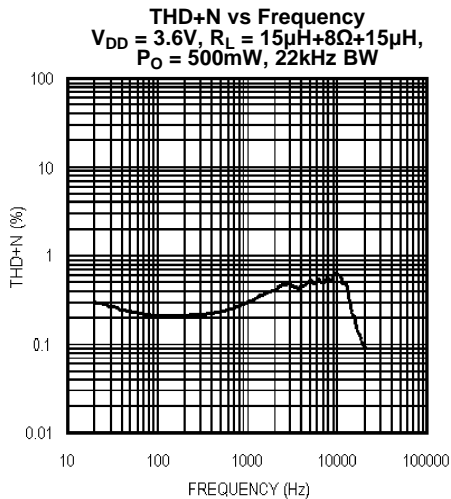


Figure 7.

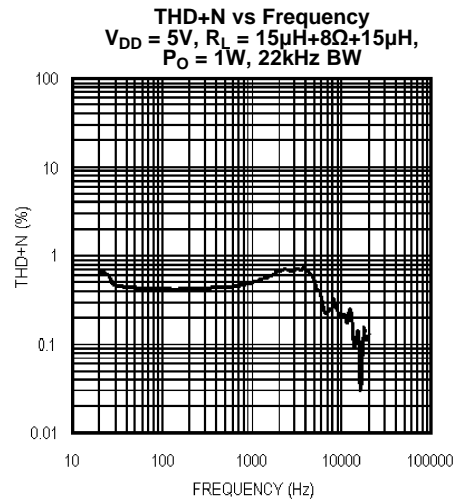


Figure 8.

Typical Performance Characteristics (continued)

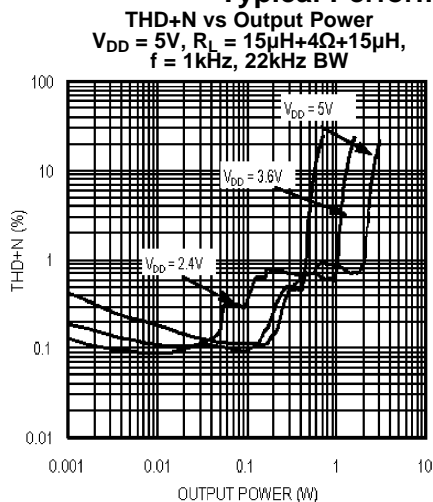


Figure 9.

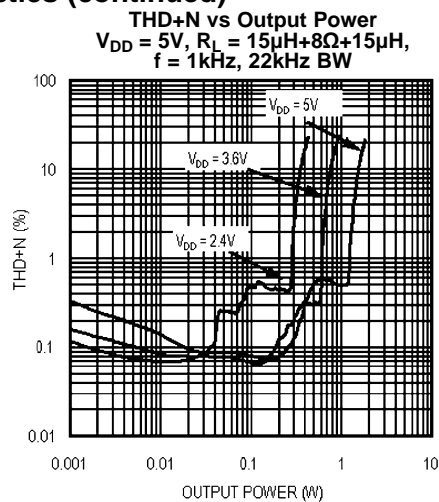


Figure 10.

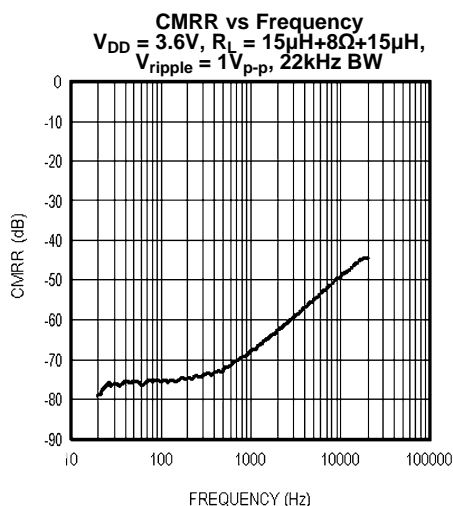


Figure 11.

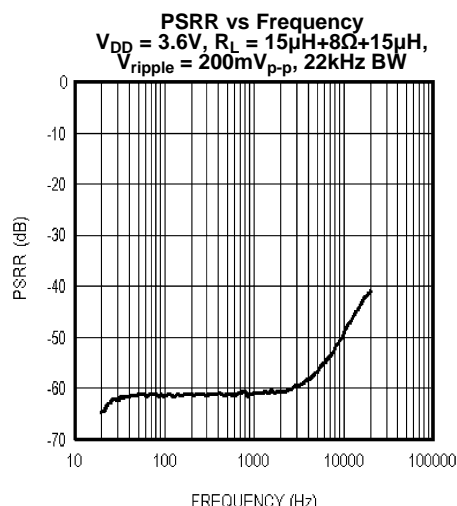


Figure 12.

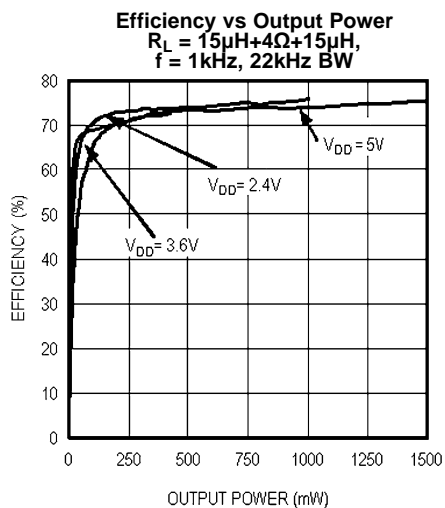


Figure 13.

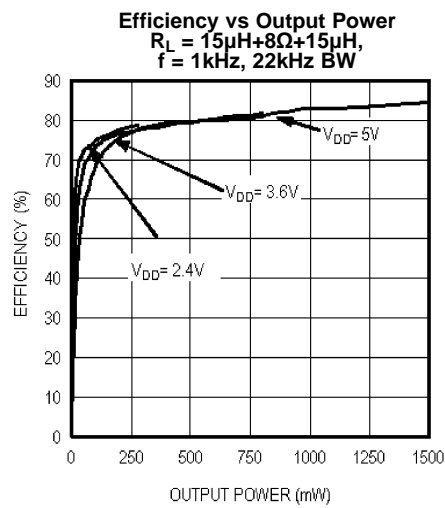


Figure 14.

Typical Performance Characteristics (continued)

Power Dissipation vs Output Power
 $R_L = 15\mu H + 4\Omega + 15\mu H$,
 $f = 1\text{kHz}, 22\text{kHz BW}$

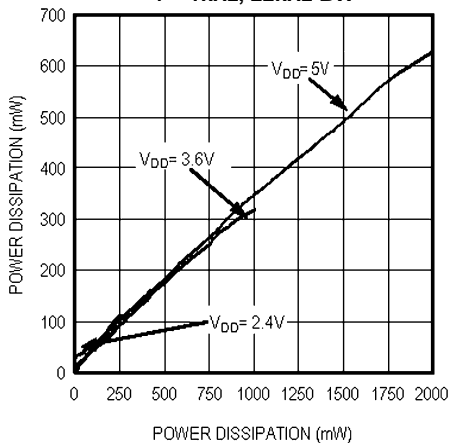


Figure 15.

Power Dissipation vs Output Power
 $R_L = 15\mu H + 8\Omega + 15\mu H$,
 $f = 1\text{kHz}, 22\text{kHz BW}$

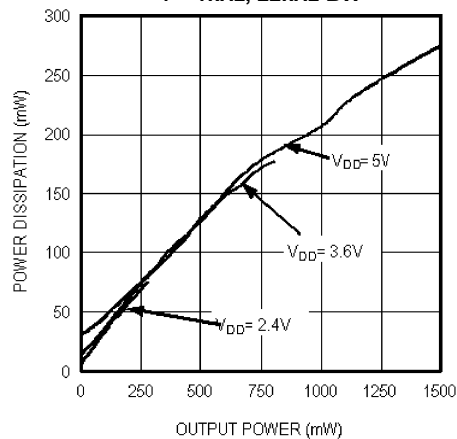


Figure 16.

Output Power vs Supply Voltage
 $R_L = 15\mu H + 4\Omega + 15\mu H$,
 $V_{DD} = 3.6\text{V}$

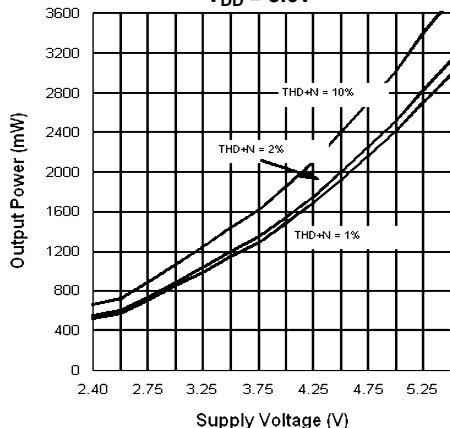


Figure 17.

Output Power vs Supply Voltage
 $R_L = 15\mu H + 8\Omega + 15\mu H$,
 $V_{DD} = 3.6\text{V}$

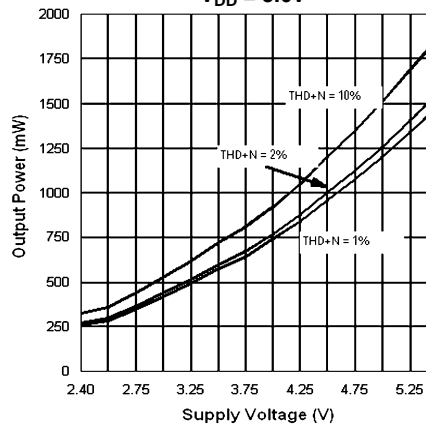


Figure 18.

Gain vs Supply Voltage
 $R_{in} = 150\text{k}\Omega$

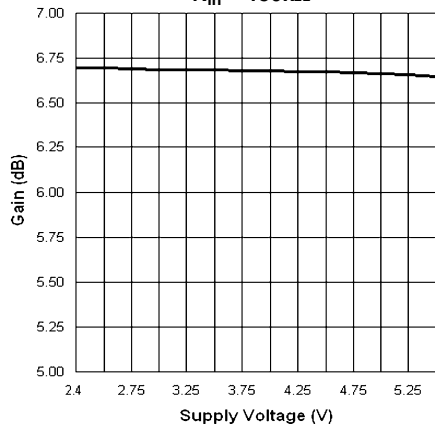


Figure 19.

Supply Current vs Supply Voltage
 $R_L = 15\mu H + 8\Omega + 15\mu H$

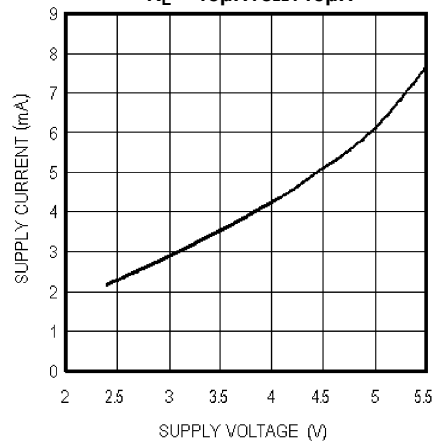


Figure 20.

Typical Performance Characteristics (continued)

Shutdown Current vs Supply Voltage
 $R_L = 15\mu\text{H} + 8\Omega + 15\mu\text{H}$

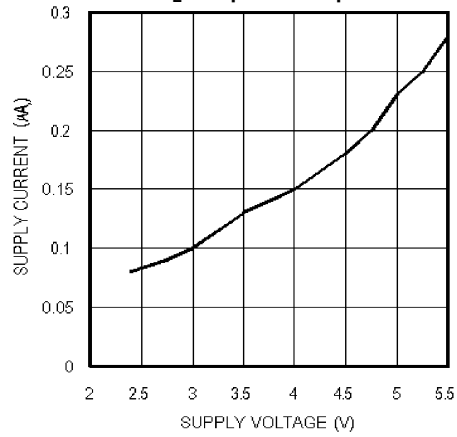


Figure 21.

APPLICATION INFORMATION

GENERAL AMPLIFIER FUNCTION

The LM4671 features a filterless modulation scheme. The differential outputs of the device switch at 300kHz from V_{DD} to GND. When there is no input signal applied, the two outputs (V_{O1} and V_{O2}) switch with a 50% duty cycle, with both outputs in phase. Because the outputs of the LM4671 are differential, the two signals cancel each other. This results in no net voltage across the speaker, thus there is no load current during an idle state, conserving power.

With an input signal applied, the duty cycle (pulse width) of the LM4671 outputs changes. For increasing output voltages, the duty cycle of V_{O1} increases, while the duty cycle of V_{O2} decreases. For decreasing output voltages, the converse occurs, the duty cycle of V_{O2} increases while the duty cycle of V_{O1} decreases. The difference between the two pulse widths yields the differential output voltage.

POWER DISSIPATION AND EFFICIENCY

In general terms, efficiency is considered to be the ratio of useful work output divided by the total energy required to produce it with the difference being the power dissipated, typically, in the IC. The key here is “useful” work. For audio systems, the energy delivered in the audible bands is considered useful including the distortion products of the input signal. Sub-sonic (DC) and super-sonic components (>22kHz) are not useful. The difference between the power flowing from the power supply and the audio band power being transduced is dissipated in the LM4671 and in the transducer load. The amount of power dissipation in the LM4671 is very low. This is because the ON resistance of the switches used to form the output waveforms is typically less than 0.25Ω . This leaves only the transducer load as a potential “sink” for the small excess of input power over audio band output power. The LM4671 dissipates only a fraction of the excess power requiring no additional PCB area or copper plane to act as a heat sink.

DIFFERENTIAL AMPLIFIER EXPLANATION

As logic supply voltages continue to shrink, designers are increasingly turning to differential analog signal handling to preserve signal to noise ratios with restricted voltage swing. The LM4671 is a fully differential amplifier that features differential input and output stages. A differential amplifier amplifies the difference between the two input signals. Traditional audio power amplifiers have typically offered only single-ended inputs resulting in a 6dB reduction in signal to noise ratio relative to differential inputs. The LM4671 also offers the possibility of DC input coupling which eliminates the two external AC coupling, DC blocking capacitors. The LM4671 can be used, however, as a single ended input amplifier while still retaining its fully differential benefits. In fact, completely unrelated signals may be placed on the input pins. The LM4671 simply amplifies the difference between the signals. A major benefit of a differential amplifier is the improved common mode rejection ratio (CMRR) over single input amplifiers. The common-mode rejection characteristic of the differential amplifier reduces sensitivity to ground offset related noise injection, especially important in high noise applications.

PCB LAYOUT CONSIDERATIONS

As output power increases, interconnect resistance (PCB traces and wires) between the amplifier, load and power supply create a voltage drop. The voltage loss on the traces between the LM4671 and the load results in lower output power and decreased efficiency. Higher trace resistance between the supply and the LM4671 has the same effect as a poorly regulated supply, increase ripple on the supply line also reducing the peak output power. The effects of residual trace resistance increases as output current increases due to higher output power, decreased load impedance or both. To maintain the highest output voltage swing and corresponding peak output power, the PCB traces that connect the output pins to the load and the supply pins to the power supply should be as wide as possible to minimize trace resistance.

The use of power and ground planes will give the best THD+N performance. While reducing trace resistance, the use of power planes also creates parasite capacitors that help to filter the power supply line.

The inductive nature of the transducer load can also result in overshoot on one or both edges, clamped by the parasitic diodes to GND and V_{DD} in each case. From an EMI standpoint, this is an aggressive waveform that can radiate or conduct to other components in the system and cause interference. It is essential to keep the power and output traces short and well shielded if possible. Use of ground planes, beads, and micro-strip layout techniques are all useful in preventing unwanted interference.

As the distance from the LM4671 and the speaker increase, the amount of EMI radiation will increase since the output wires or traces acting as antenna become more efficient with length. What is acceptable EMI is highly application specific. Ferrite chip inductors placed close to the LM4671 may be needed to reduce EMI radiation. The value of the ferrite chip is very application specific.

POWER SUPPLY BYPASSING

As with any power amplifier, proper supply bypassing is critical for low noise performance and high power supply rejection ratio (PSRR). The capacitor (C_S) location should be as close as possible to the LM4671. Typical applications employ a voltage regulator with a 10 μ F and a 0.1 μ F bypass capacitors that increase supply stability. These capacitors do not eliminate the need for bypassing on the supply pin of the LM4671. A 1 μ F tantalum capacitor is recommended.

SHUTDOWN FUNCTION

In order to reduce power consumption while not in use, the LM4671 contains shutdown circuitry that reduces current draw to less than 0.01 μ A. The trigger point for shutdown is shown as a typical value in the [Electrical Characteristics](#) Tables and in the Shutdown Hysteresis Voltage graphs found in the [Typical Performance Characteristics](#) section. It is best to switch between ground and supply for minimum current usage while in the shutdown state. While the LM4671 may be disabled with shutdown voltages in between ground and supply, the idle current will be greater than the typical 0.01 μ A value.

The LM4671 has an internal resistor connected between GND and Shutdown pins. The purpose of this resistor is to eliminate any unwanted state changes when the Shutdown pin is floating. The LM4671 will enter the shutdown state when the Shutdown pin is left floating or if not floating, when the shutdown voltage has crossed the threshold. To minimize the supply current while in the shutdown state, the Shutdown pin should be driven to GND or left floating. If the Shutdown pin is not driven to GND, the amount of additional resistor current due to the internal shutdown resistor can be found by [Equation \(1\)](#) below.

$$(V_{SD} - GND) / 60k\Omega \quad (1)$$

With only a 0.5V difference, an additional 8.3 μ A of current will be drawn while in the shutdown state.

PROPER SELECTION OF EXTERNAL COMPONENTS

The gain of the LM4671 is set by the external resistors, R_i in [Figure 1](#), The Gain is given by [Equation \(2\)](#) below. Best THD+N performance is achieved with a gain of 2V/V (6dB).

$$A_v = 2 * 150 k\Omega / R_i (V/V) \quad (2)$$

It is recommended that resistors with 1% tolerance or better be used to set the gain of the LM4671. The R_i resistors should be placed close to the input pins of the LM4671. Keeping the input traces close to each other and of the same length in a high noise environment will aid in noise rejection due to the good CMRR of the LM4671. Noise coupled onto input traces which are physically close to each other will be common mode and easily rejected by the LM4671.

Input capacitors may be needed for some applications or when the source is single-ended (see [Figures 3](#) and [5](#)). Input capacitors are needed to block any DC voltage at the source so that the DC voltage seen between the input terminals of the LM4671 is 0V. Input capacitors create a high-pass filter with the input resistors, R_i . The –3dB point of the high-pass filter is found using [Equation \(3\)](#) below.

$$f_c = 1 / (2\pi R_i C_i) (Hz) \quad (3)$$

The input capacitors may also be used to remove low audio frequencies. Small speakers cannot reproduce low bass frequencies so filtering may be desired. When the LM4671 is using a single-ended source, power supply noise on the ground is seen as an input signal by the +IN input pin that is capacitor coupled to ground (see [Figures 5 -7](#)). Setting the high-pass filter point above the power supply noise frequencies, 217Hz in a GSM phone, for example, will filter out this noise so it is not amplified and heard on the output. Capacitors with a tolerance of 10% or better are recommended for impedance matching.

DIFFERENTIAL CIRCUIT CONFIGURATIONS

The LM4671 can be used in many different circuit configurations. The simplest and best performing is the DC coupled, differential input configuration shown in [Figure 22](#). [Equation \(2\)](#) above is used to determine the value of the R_i resistors for a desired gain.

Input capacitors can be used in a differential configuration as shown in [Figure 23](#). [Equation \(3\)](#) above is used to determine the value of the C_i capacitors for a desired frequency response due to the high-pass filter created by C_i and R_i . [Equation \(2\)](#) above is used to determine the value of the R_i resistors for a desired gain.

The LM4671 can be used to amplify more than one audio source. [Figure 24](#) shows a dual differential input configuration. The gain for each input can be independently set for maximum design flexibility using the R_i resistors for each input and [Equation \(2\)](#). Input capacitors can be used with one or more sources as well to have different frequency responses depending on the source or if a DC voltage needs to be blocked from a source.

SINGLE-ENDED CIRCUIT CONFIGURATIONS

The LM4671 can also be used with single-ended sources but input capacitors will be needed to block any DC at the input terminals. [Figure 25](#) shows the typical single-ended application configuration. The equations for Gain, [Equation \(2\)](#), and frequency response, [Equation \(3\)](#), hold for the single-ended configuration as shown in [Figure 25](#).

When using more than one single-ended source as shown in [Figure 26](#), the impedance seen from each input terminal should be equal. To find the correct values for C_{i3} and R_{i3} connected to the +IN input pin the equivalent impedance of all the single-ended sources are calculated. The single-ended sources are in parallel to each other. The equivalent capacitor and resistor, C_{i3} and R_{i3} , are found by calculating the parallel combination of all C_i values and then all R_i values. [Equations \(4\)](#) and [\(5\)](#) below are for any number of single-ended sources.

$$C_{i3} = C_{i1} + C_{i2} + C_{in} \dots (\mu\text{F}) \quad (4)$$

$$R_{i3} = 1 / (1/R_{i1} + 1/R_{i2} + 1/R_{in} \dots) (\Omega) \quad (5)$$

The LM4671 may also use a combination of single-ended and differential sources. A typical application with one single-ended source and one differential source is shown in [Figure 27](#). Using the principle of superposition, the external component values can be determined with the above equations corresponding to the configuration.

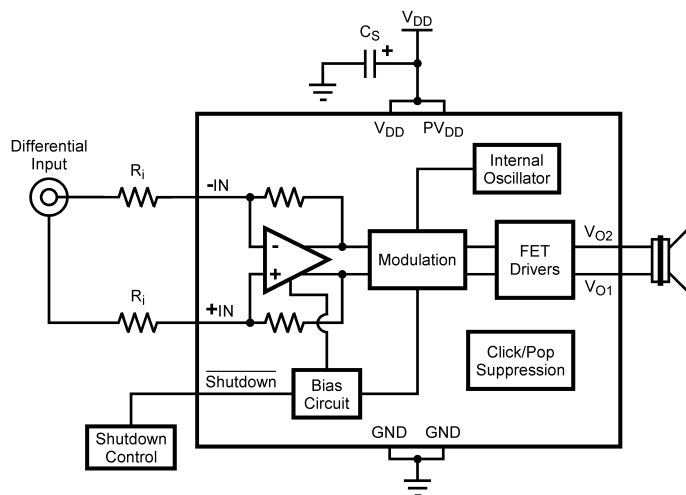


Figure 22. Differential input configuration

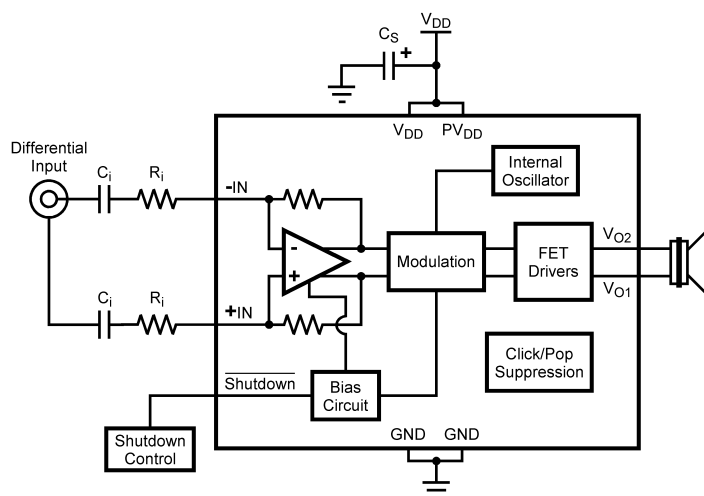


Figure 23. Differential input configuration with input capacitors

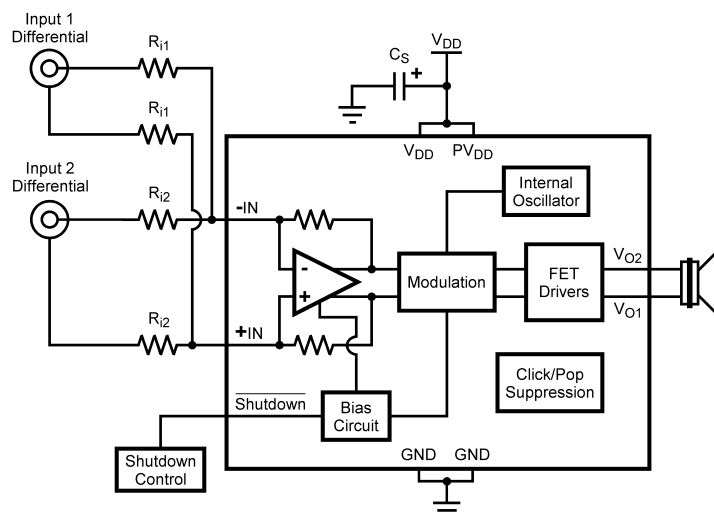


Figure 24. Dual differential input configuration

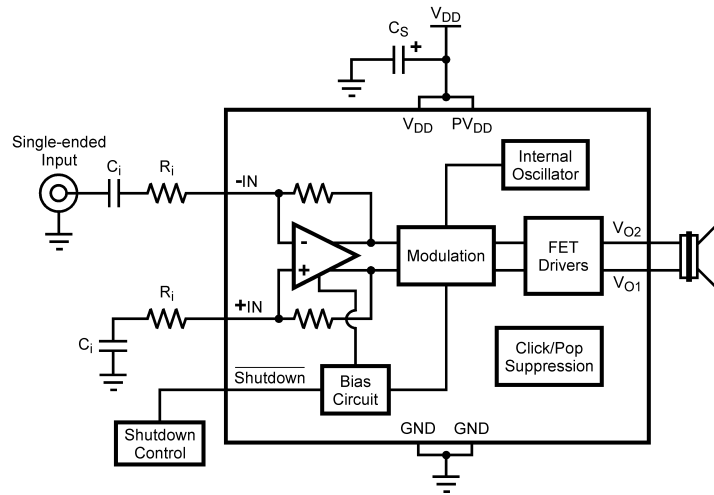


Figure 25. Single-ended input configuration

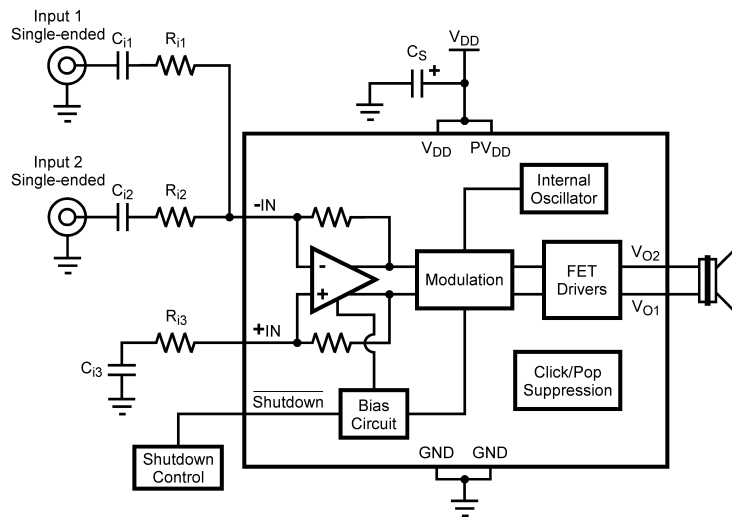


Figure 26. Dual single-ended input configuration

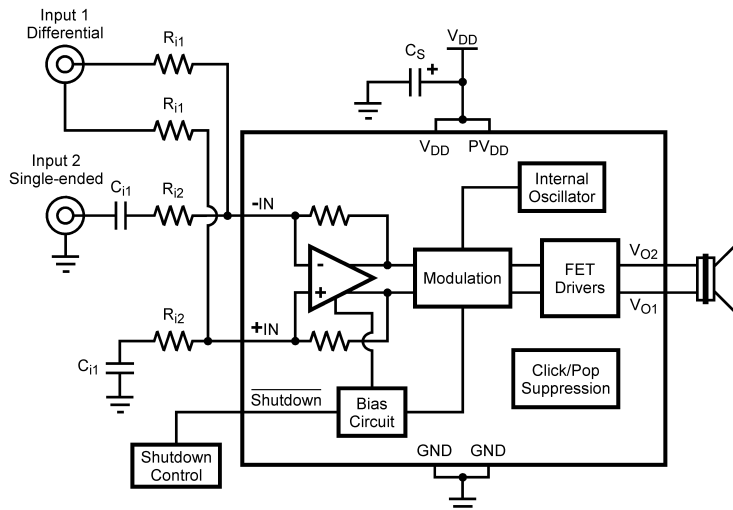


Figure 27. Dual input with a single-ended input and a differential input

REFERENCE DESIGN BOARD SCHEMATIC

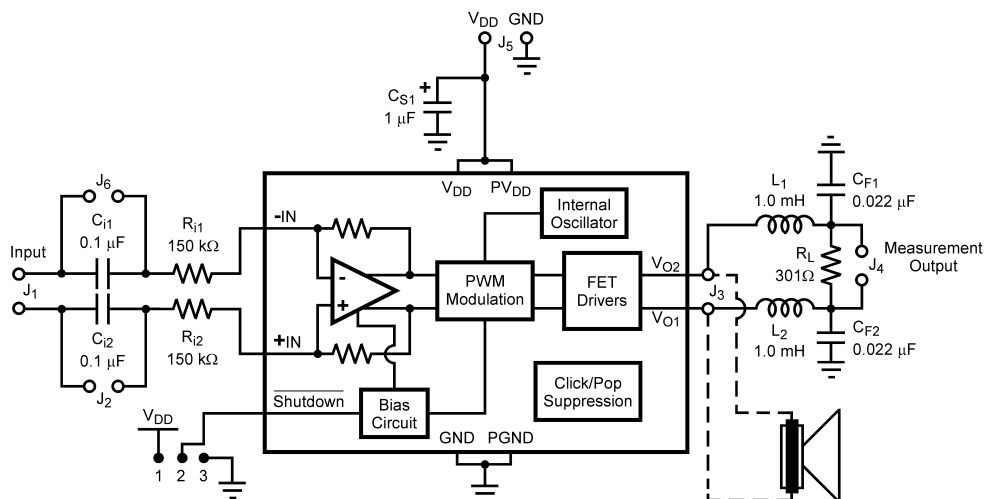


Figure 28. Reference Design Board Schematic

In addition to the minimal parts required for the application circuit, a measurement filter is provided on the evaluation circuit board so that conventional audio measurements can be conveniently made without additional equipment. This is a balanced input, grounded differential output low pass filter with a 3dB frequency of approximately 35kHz and an on board termination resistor of 300Ω (see Figure 28). Note that the capacitive load elements are returned to ground. This is not optimal for common mode rejection purposes, but due to the independent pulse format at each output there is a significant amount of high frequency common mode component on the outputs. The grounded capacitive filter elements attenuate this component at the board to reduce the high frequency CMRR requirement placed on the analysis instruments.

Even with the grounded filter the audio signal is still differential, necessitating a differential input on any analysis instrument connected to it. Most lab instruments that feature BNC connectors on their inputs are **NOT** differential responding because the ring of the BNC is usually grounded.

The commonly used Audio Precision analyzer is differential, but its ability to accurately reject high frequency signals is questionable necessitating the on board measurement filter. When in doubt or when the signal needs to be single-ended, use an audio signal transformer to convert the differential output to a single ended output. Depending on the audio transformer's characteristics, there may be some attenuation of the audio signal which needs to be taken into account for correct measurement of performance.

Measurements made at the output of the measurement filter suffer attenuation relative to the primary, unfiltered outputs even at audio frequencies. This is due to the resistance of the inductors interacting with the termination resistor (300Ω) and is typically about -0.25dB (3%). In other words, the voltage levels (and corresponding power levels) indicated through the measurement filter are slightly lower than those that actually occur at the load placed on the unfiltered outputs. This small loss in the filter for measurement gives a lower output power reading than what is really occurring on the unfiltered outputs and its load.

LM4671 DSBGA BOARD ARTWORK

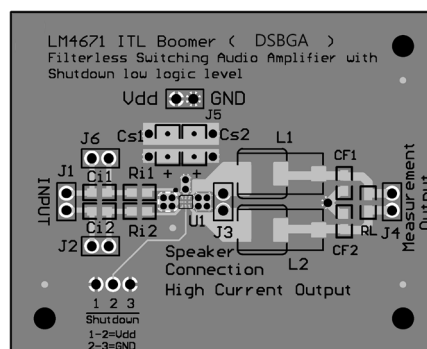


Figure 29. Composite View

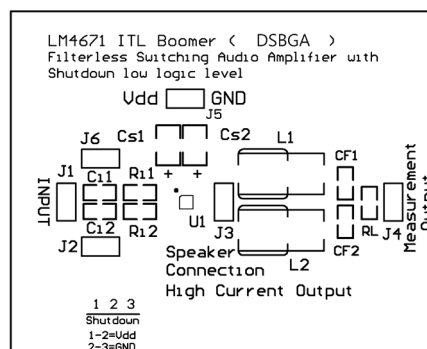


Figure 30. Silk Screen

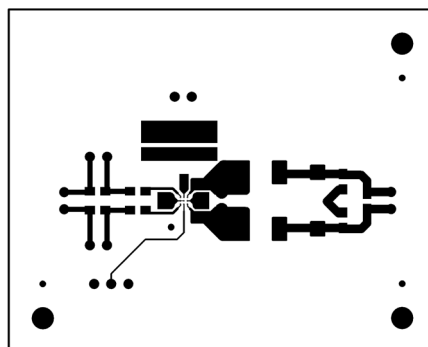


Figure 31. Top Layer

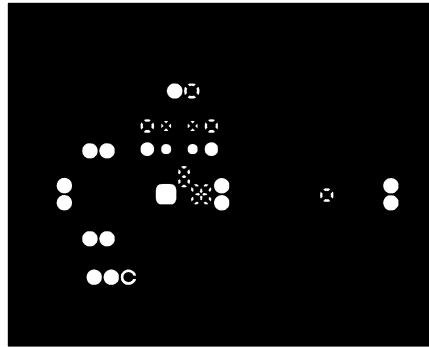


Figure 32. Internal Layer 1, GND

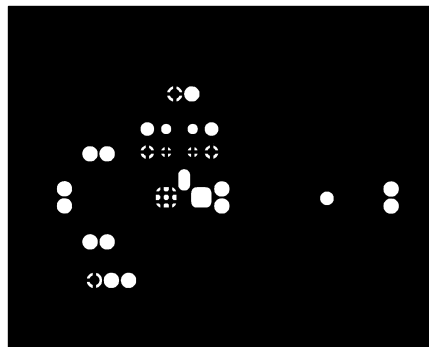


Figure 33. Internal Layer 2, GND

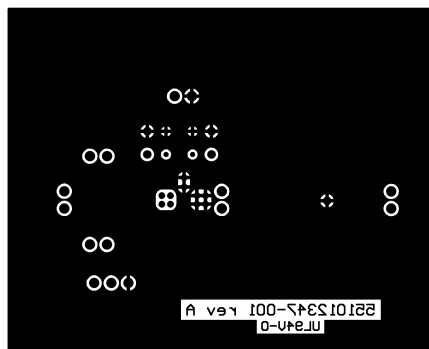




Figure 34. Bottom Layer

Revision History

Rev	Date	Description
1.0	03/16/05	Initial release.
1.01	12/17/07	Some text edits.
1.02	03/03/08	Edited the SHUTDOWN FUNCTION section (under Application Information).

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LM4671ITL/NOPB	ACTIVE	DSBGA	YZR	9	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	G E7	
LM4671ITLX/NOPB	ACTIVE	DSBGA	YZR	9	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	G E7	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

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QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM4671ITL/NOPB	DSBGA	YZR	9	250	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LM4671ITLX/NOPB	DSBGA	YZR	9	3000	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1

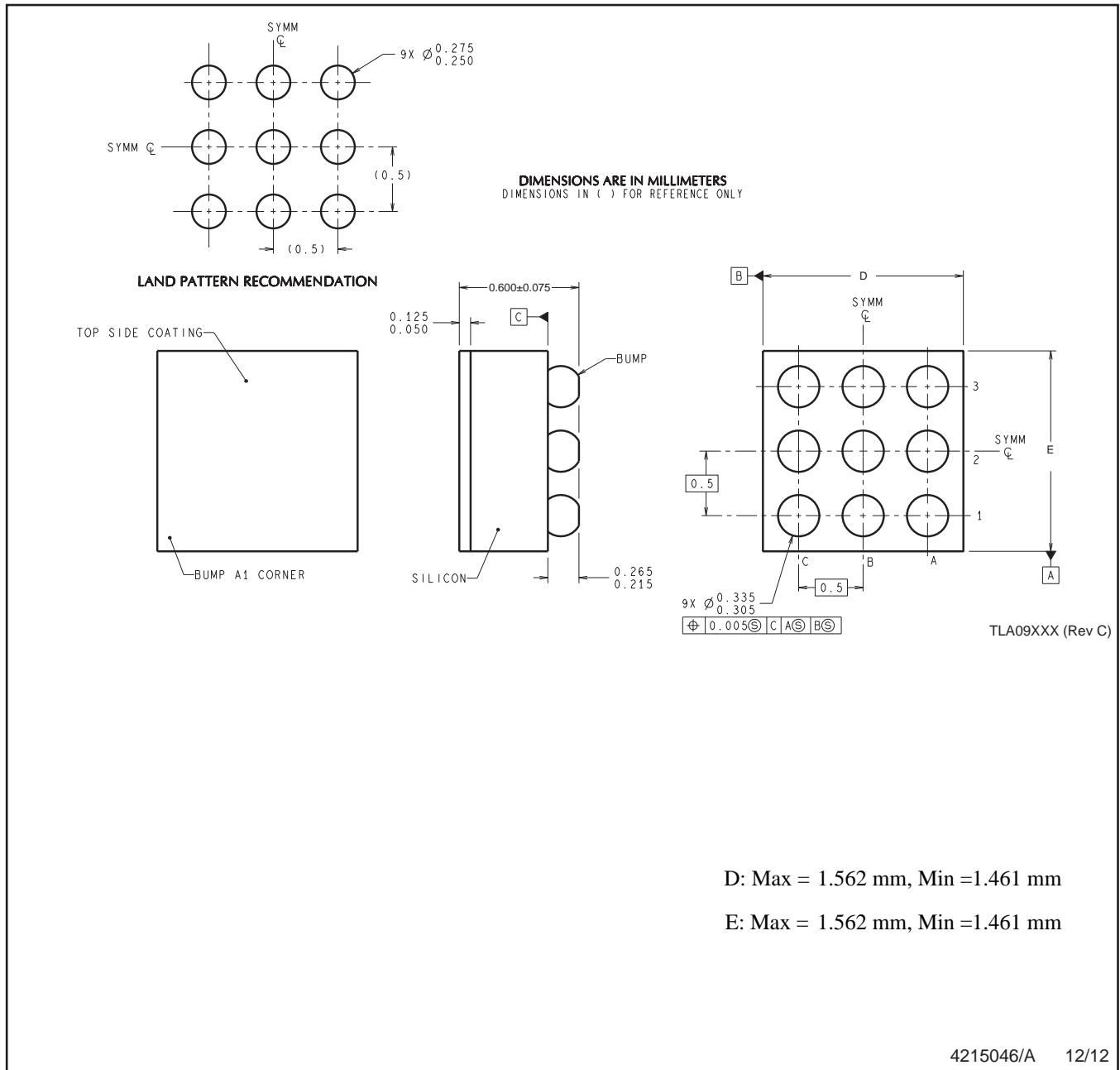
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM4671ITL/NOPB	DSBGA	YZR	9	250	203.0	190.0	41.0
LM4671ITLX/NOPB	DSBGA	YZR	9	3000	206.0	191.0	90.0

YZR0009



D: Max = 1.562 mm, Min = 1.461 mm

E: Max = 1.562 mm, Min = 1.461 mm

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
B. This drawing is subject to change without notice.

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