LB1980



3-Phase Brushless Motor Driver for VCR Capstan Motors

Overview

The LB1980 is a 3-phase brushless motor driver IC that is particularly well suited for driving VCR capstan motors.

Functions

- Three-phase full wave current linear drive
- Torque ripple correction circuit with variable correction ratio
- Current limiter circuit
- Output stage up/down oversaturation protection circuit (no external capacitor required)
- FG amplifier
- Thermal shutdown circuit



Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
	V _{CC} max		7	٧
Maximum supply voltage	V _S max		24	V
Maximum output current	lo max		1.3	А
	0.1	With an arbitrarily large heat sink	12.5	W
Allowable power dissipation Pd r		Independent IC	0.77	W
Operating temperature	Topr		-20 to +75	°C
Storage temperature	Tstg		55 to +150	۴C

Allowable Operating Ranges at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Quantum theme	٧s		5 to 22	V
Supply voltage	V _{CC}		4.5 to 5.5	v
Hall input amplitude	VHALL	Between Hall inputs	±30 to ±80	mVo-p
GSENSE input range	VGSENSE	With respect to the control system ground	-0.20 to +0.20	V

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Package Dimensions

unit: mm

3206A-QFP34HA



			Ratings			Unit	
Parameter	Symbol	Conditions	min	typ	max		
V _{CC} current drain	lcc	R _L = ∞, V _{CTL} = 0 V, V _{LIM} = 0 V(When stopped)	. –	12	18	mA	
[Output]	.	· · · · · · · · · · · · · · · · · · ·		·····			
	Vo(sat)1	$ I_O = 500 \text{ mA}, \text{ Rf} = 0.5 \Omega, \text{ Sink + Source} \\ V_{CTL} = V_{LIM} \approx 5 \text{ V(With saturation protection)} $	-	2.1	2.6	v	
	Vo(sat)2	$I_O = 1.0 \text{ A}, \text{ Rf} = 0.5 \Omega, \text{ Sink + Source}$ $V_{CTL} = V_{LIM} = 5 V(With saturation protection)$	-	2.6	3.5	v	
Output leakage current	lo leak			-	1.0	mA	
[FR]							
FR pin input threshold voltage	V _{FSR}		2.25	2.50	2.75	V	
FR pin input bias current	lb(FSR)		5.0	-	-	μA	
[Control]							
CTLREF pin voltage	VCREF		2.05	2.15	2.25	v	
CTLREF pin input range	VCREFIN		1.50	-	3.50	V	
CTL pin input bias current	Ib(CTL)	V _{CTL} = 5 V, with CTLREF open	-	-	4.0	μA	
CTL control start voltage	VCTL(ST)	With Rf = 0.5 Ω , V _{LIM} = 5 V, I _O ≥ 10 mA With the Hall input logic fixed(u, v, w = H, H, L)	2.00	2.15	2.30	v	
CTL pin control Gm	Gm(CTL)	With Rf = 0.5Ω , $\Delta I_0 = 200 \text{ mA}$ With the Hall input logic fixed(u, v, w = H, H, L)	0.46	0.58	0.70	AV	
[Current Limiter]	·				. <u> </u>		
LIM current limiter offset voltage	Volf(LIM)	With Rf = 0.5 Ω , V _{CTL} = 5 V, I _Q ≥ 10 mA With the Hall input logic fixed(u, v, w = H, H, L)	140	200	260	mV	
LIM pin Input blas current	Ib(LIM)	VCTL = 5 V, CTLREF: OPEN, VLIM = 0 V	-2.5		_	μA	
LIM pin current limit level	llim	With Rf = 0.5Ω , V _{CTL} = 5 V, V _{LIM} = 20.8 V With the Hall input logic fixed(u, v, w = H, H, L)	830	900	970	mA	
[Hall Amplifies]							
Hall amplifier input offset voltage	Voff(HALL)		-6	+	+6	mV	
Half amplifier input blas current	Ib(HALL)		-	1.0	3.0	μA	
Hall amplifier common-mode input voltage	Vcm(HALL)		1.3	~	3.3	v	
[TRC]							
Torque ripple correction ratio	TRC	For the troughs and peaks in the Rf waveform at $I_0 = 200 \text{ mA}(\text{Rf} = 0.5 \Omega, \text{ADJ}; \text{open})(\text{note 1})$	-	9	_	%	
ADJ pin voltage	Vadj		2.37	2.50	2.63	Ý I	
[FG Amplifier]							
FG amplifier input offset voltage	Voff(FG)		-8	-	+8	mV	
FG amplifier input bias current	lb(FG)		100	-	_	nA	
FG amplitier output saturation vollage	Vo(sat)(FG)	For the sink side and the internal pull-up resistor load	-	-	0.5	×	
FG amplifier voltage gain	VG(FG)	For open loop at f = 10 kHz	41.5	44.5	47.5	dB	
FG amplifier common-mode input voltage	VCM(FG)		0.5	-	4.0	v	
[Saturation]							
Saturation prevention circuit lower side voltage setting	Vo(sat)(DET)	$I_O = 10$ mA, RI = 0.5 Ω The voltage between each OUT and RI when V _{CTL} = V _{LIM} = 5 V.	0.175	0.25	0.325	v	
[TSD]							
TSD operating temperature	T-TSD	(Design target value)*	-	180		°C	
TSD temperature hysteresis	ΔTSD	(Design target value)*	-	20	_	°C	
	•	······································		<u> </u>			

Notes: *Items marked with an asterisk are design target values and are not measured. 1. The torque ripple correction ratio is determined from the Rf voltage waveform as follows.



 $=\frac{2\cdot(Vp-Vb)}{Vp+Vb}\cdot 100\,(\%)$

Pin Assignment



Note: For ground potential stabilization, the FRAME pins must be connected to ground.

Truth Table and Control Functions

	Source Sink	Hall input			FR	
		U	v	W] [
1	$V \rightarrow W$				Н	
	$W \rightarrow V$				L	
2	$U \rightarrow W$				н	
	$W \rightarrow U$		L.	L	L	
3	U -→ V		L	н	н	
	$V \rightarrow U$				L	
4	$W \rightarrow V$		L	н	Н	
	$V \rightarrow W$				L	
5	W → U			н	н	
	U→W		l H		L	
6	$V \rightarrow U$				н	
	$U \rightarrow V$				L	

Note: An H in the FR column indicates a voltage of 2.75 V or higher, and an L indicates a voltage of 2.25 V or lower. (When V_{CC} = 5 V.)

Note: For the Hall inputs, H indicates that the + input is 0.01 V or more higher than the input, and L indicates that the + input is 0.01 V or more lower than the - input, for all phases.

Note: Since this drive technique is a 180° conducting technique, the phases other than the sink and source phases are never off.

[Control Function and Current Limiter Function]

2.10Vtyp

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Output current, IoUT

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Control Characteristics

Gm⇔0,58A/Ytyp

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Control voltage, VCTL

VLIM≂5V



Pin Functions

Pin No.	Pin name	Pin functions	Equivalent circuit
1	FR	Forward/reverse selection pin.	
		Forward or reverse is selected by the voltage applied to this pin.	
r .		(Vth = 2.5 V (typical) at V_{CC} = 5 V)	VCC VCC
23	ADJ	This pin allows the torque ripple correction ratio to be adjusted	
		externally. Applications that adjust the correction ratio should	
		apply a voltage to the ADJ pin with a low-impedance circuit.	
		The correction ratio falls as the applied voltage increases, and	
		increases as the applied voltage decreases. The correction ratio	★
		can be varied over a range from about 0 to twice the ratio when	‴‴ [≭] [‡] ₩ ★ [
		this pin is left open. (This voltage is set internally to about $V_{CC}/2$,	
	· ·	and the input impedance is about 5 k Ω .)	DEGONA
2	GND	Ground for circuits other than the output transistors	
		The output transistor minimum potential is that of the resistor Rf.	
5	FGin-	When the FG amplifier is used as an inverting input amplifier,	
		a feedback resistor must be inserted between FG-OUT	Acc V
		and this pin.	
			E.A
			<u>م</u> مر ا
6	FGin+	This is the noninverting input when the FG amplifier is used as	
		a differential input amplifier. There is no internally applied bias.	
			🛉 🛉
			777
			¥08021
8	FG-OUT	This is the FG amplifier output. It has an internal load resistance.	
			5 5260
		•	
12	FC	Frequency characteristics correction for the speed control loop	
			10605
1			
9	CTL	Speed control connection. Speed control is implemented as a	······
		fixed current drive in which current feedback is applied from Rf.	
		Gm = 0.58 A/V (typical) at Rf = 0.5 Ω	
10	CTLREF	Control reference. Although this pin is set to about 0.43 * V_{CC}	
		internally, it can be changed by applying a voltage with	
		a low-impedance circuit. (It has an input impedance of about	
		4.3 kΩ (typical).)	
11	LIM	Current limiter function control. The voltage applied to this pin	1 th th th th
		changes the output current linearly. Slope = 0.5 A/V (typical)	ECG90*
		at $Rf = 0.5 \ \Omega$.	
13	Uin+	U phase Hall element input. IN+ must be higher than IN-	Each (+) input Each (-) input
14	Uln-	for a logical high input level.	
15	Vin+	V phase Hall element input. IN+ must be higher than IN-	
16	Vin-	for a logical high input level.	
17	Win+	W phase Hall element input. IN+ must be higher than IN-	(¹⁰⁰ ** ⁷⁷
18	Win-	for a logical high input level.	777 AD5053
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Pin No.	Pin name	Pin functions	Equivalent circuit
19	V _{CC}	Power supply for all IC internal circuits except for the output block. This voltage must be stabilized so that ripple and noise do not enter the IC circuits.	
22	Vs	Output block power supply	
24	Rf (PWR)	These are the output current detection pins, and the control block current feedback is applied by connecting the resistor	
33	Rf (SENSE)	Rf between these pins and ground. Additionally, the lower side oversaturation protection circuit and the torque ripple correction circuit operate according to the voltages on these pins. In particular, since the oversaturation protection level is set by the voltages on these pins, if the value of Rf is lowered excessively, the effectiveness of the lower side oversaturation protection in the high-current region will be degraded. Note that the POWER and SENSE pins must be connected together.	VS VS Lower side saturation prevention circuit input block VCC VCC VS VS VS Lower side saturation prevention circuit input block VCC VCC VS VS VS VS VS VS VS VS VS VS VS VS VS
27	U-OUT	U phase output. (A spark killer diode is built in.)	A08034
29	ν-ουτ	V phase output. (A spark killer diode is built in.)	
31	W-QUT	W phase output. (A spark killer diode Is built in.)	
34	GSENSE	Ground sense. By connecting this pin near the ground for the Rf resistor side of the motor ground wiring that includes Rf, the influence that the common ground Impedance has on Rf can be removed. (This pin must not be left open.)	

Block Diagram





Sample Application Circuit

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